

# Joint PDF Tracking Control for a Class of Multivariate Time-varying Stochastic Descriptor Systems<sup>\*</sup>

Lei Guo<sup>\*</sup> Liping Yin<sup>\*\*</sup>

<sup>\*</sup> School of Instrument Science and Opto-Electronics Engineering,  
Beihang University, Beijing, China (e-mail: l.guo@seu.edu.cn).

<sup>\*\*</sup> Research Institute of Automation, Southeast University, Nanjing,  
China (Tel: 86-25-83795609 ; e-mail: yinlipingqufu@163.com).

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**Abstract:** This paper considers a new tracking control problem for a class of nonlinear stochastic descriptor systems, where the tracked target is a given joint probability density function (JPDF). The controlled plants can be represented by multivariate discrete-time descriptor systems with non-Gaussian disturbances and nonlinear output equations. The control objective is to find crisp algorithms such that the conditional output JPDFs can follow the given target JPDF. Rather than using statistic methods such as Bayesian estimation or Monte Carlo methods, we establish a direct relationship between the JPDFs of the transformed tracking error and the stochastic input. An optimization approach is applied to present recursive algorithms such that the distances between the output distributions and the desired one are minimized. Furthermore, a stabilization suboptimal control strategy is proposed by using of LMI-based Lyapunov theory. Simulations are provided to demonstrate the effectiveness of the stochastic tracking control algorithms.

Keywords: probability density function; descriptor system; tracking; stochastic control; optimal control; Lyapunov theory; simulation of stochastic system.

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## 1. INTRODUCTION

The control problems for descriptor systems (or generalized systems, singular systems) have drawn considerable attention of many researchers due to their extensive applications in engineering. Many new feasible control approaches based on Lyapunov theory have extensively been presented for linear descriptor systems (see *e. g.* (Dai89, GM03, HL99, Tak98, WM94)). For stochastic descriptor systems, most of the existing results focused on generalized Kalman filtering theory, where linear plants with Gaussian noises were concerned (see *e. g.* (NCF99, YLM96, ZXS99)). It is shown that in the presence of descriptor (implicit) dynamics, the Kalman filtering problem for linear descriptor systems becomes much more complex than that for conventional systems.

On the other hand, for the conventional stochastic systems without descriptor dynamics, stochastic processes and stochastic control have been widely studied theoretically and applied in practice in the past decades. However, in most existing works only linear stochastic

systems with Gaussian random variables were considered, and the control performance objectives were confined to be either expectation or variance of the stochastic output (see *e. g.* (Astrom70, Mao02, Pap91, Pet00, UP01, Wang00)). It has been shown that in many practical processes such as paper and board making systems, the related stochastic signals are non-Gaussian and/or the plants have strong non-linearity (see *e.g.* (Wang00)). For these stochastic systems with non-Gaussian inputs or outputs, the expectation and variance are not sufficient to characterize the statistics of the concerned random vectors. Generally speaking, for descriptor systems with non-Gaussian stochastic noises and nonlinear dynamics, few control and filtering applicable results have been obtained up to date.

For the past few years, new control strategies for the shape of the output probability density functions (PDF) have been developed for the conventional (non-descriptor) stochastic systems, where the stochastic variables are not confined to be only Gaussian and the control objective is the shape of output (see *e.g.* (Kar96, Wang00, Wang02)). Up to now, mainly there are two kinds of PDF control strategies having been proposed for some conventional stochastic

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systems. One concentrated on tuning the dynamic weights which correspond to the output PDFs after modeling the dynamics between the control input and the output PDFs using B-spline expansions (see *e. g.* (Wang00, Wang02)), another attempted to establish the relationships between the PDFs of the stochastic input and output, and made use of various optimization approaches (see, *e.g.* (Wang03)). However, most existing results on PDF control and minimum entropy control only considered single-input-single-output (SISO) conventional systems. It is noted that multivariate stochastic processes are much more complicated where in general some new methodologies involved in conditional joint probability density functions (JPDFs) should be discussed (see, *e.g.* (GW03)). Moreover, since stability can not be guaranteed with most of the proposed algorithms, stability analysis turns to be another obstacle in applications of these approaches.

In this paper, new PDF tracking control strategies are developed for a group of multivariate time-varying descriptor systems with non-Gaussian stochastic inputs. The control objective is to find crisp recursive design procedures such that the system outputs, which generally are non-Gaussian random vectors, can follow a target vector corresponding to a given joint distribution. After a new relationship is presented between the JPDFs of the tracking errors and inputs of the descriptor systems, the suboptimal control strategies are constructed based on gradient algorithms. Furthermore, a recursive stabilization suboptimal control law is proposed with which the stability of the closed loop system can be guaranteed. Finally simulations are provided to demonstrate the effectiveness of the proposed approaches.

## 2. PROBLEM FORMULATION

### 2.1 Process Model and Its transformation

We firstly consider the time-varying descriptor system described by

$$\Sigma_D : \begin{cases} E x_{k+1} = A_k x_k + B_{1k} u_k + B_{2k} w_k \\ y_k = h(x_k, v_k, r) \end{cases} \quad (1)$$

where  $x_k \in R^n$  is the state sequence,  $y_k \in R^l$  is the controlled output sequence,  $u_k \in R^p$  is the control sequence,  $w_k \in R^q$  and  $v_k \in R^l$  are the stochastic disturbance sequences, and  $r$  is the known reference input.  $E$ ,  $A_k$  and  $B_{ik}$  ( $i = 1, 2$ ) are known time-varying matrices, where  $E$  satisfying  $rank(E) = r < n$ . It should be noted that

the concerned plants can be generalized to more extensive nonlinear descriptor systems under some appropriate assumptions (see also *Remark 3*).

**Definition 1.** For a discrete-time linear descriptor system, a pair  $(E, A)$  is called regular if  $\det(sE - A)$  is not identically zero and impulse-free if the degree of  $\det(sE - A)$  is equal to  $rank(E)$ .

Similarly to most existing papers on descriptor systems, the following condition for admissibility is needed in this paper (see *e. g.* (HL99, Tak98)).

**Assumption 1.**  $(E, A)$  in  $\Sigma_D$  is regular and impulse-free at every sample time.

On the other hand, the involved stochastic disturbances  $w_k$  and  $v_k$  are supposed to be non-Gaussian but with known JPDFs. The following assumption is also quite general for the stochastic inputs, which have been shown to exist in many practical processes such as paper making systems (see *e.g.* (Wang00)).

**Assumption 2.** The random vectors  $w_k, v_k$  ( $k = 0, 1, 2, \dots$ ) (and its elements) are bounded, continuous, independent mutually and with the known JPDFs  $\gamma_w(\tau)$  and  $\gamma_v(\tau)$  defined on  $[a, b]^q$  and  $[a, b]^l$ , respectively.

The nonlinear function is required to satisfy the following condition.

**Assumption 3.**  $h(\cdot, \cdot, \cdot)$  are known Borel measurable and smooth nonlinear functions of their arguments, and  $h(0, 0, 0) = 0$ .

At each sample time  $k$ , random output variable  $y_k$  can be characterized by its JPDF  $\gamma_{y_k}(\tau)$  defined on  $[\alpha, \beta]^l$ . Generally,  $y_k$  is non-Gaussian random vector even if  $w_k$  and  $v_k$  are Gaussian ones due to the existence of nonlinearity.

For any  $\Sigma_D$  satisfying *Assumption A.1*, there exist invertible matrices  $M$  and  $N$  such that

$$M^T E N = \begin{bmatrix} I_r & 0 \\ 0 & 0 \end{bmatrix}, \quad M^T B_{2k} = \begin{bmatrix} \tilde{B}_{21k} \\ 0 \end{bmatrix}, \quad (2)$$

where  $x_k$  can be divided as  $x_{ik} \in R^{n_i}$ ,  $i = 1, 2$  with  $n_1 + n_2 = n$ , correspondingly. Thus, we can get the following result.

**Lemma 1** If  $(E, A)$  is regular and impulse-free, then  $\Sigma_D$  can be transformed to be  $\Sigma_D^s$ , which is described by

$$\Sigma_D^s : \begin{cases} x_{1k+1} = A_{1k} x_{1k} + B_{11k} u_k + B_{21k} w_k \\ x_{2k} = A_{2k} x_{1k} + B_{12k} u_k \\ y_k = H(x_{1k}, u_k, w_k, v_k, r) \end{cases} \quad (3)$$

**Proof.** This proof is similar to Theorem 1 of (HL99) or (Tak98).

### 2.2 Problem formulation and calculations of JPPDFs

In this paper, the tracking problem is considered for the multivariate stochastic descriptor systems described by (1) or (3). Different from a reference vector in classical tracking control theory, the tracked target is supposed to be a reference JPPDF, which is denoted by  $\gamma_d(\tau)$ .

The control objective is to construct  $u_k$  such that the conditional JPPDFs  $\gamma_{y_k}(\tau|x_k, u_k)$ , which is also a function of system output  $y_k$  and reference input  $r$ , satisfying

$$\gamma_{y_k}(\tau|x_k, u_k) \rightarrow \gamma_d(\tau), k \rightarrow +\infty.$$

Since the distance between  $\gamma_{y_k}(\tau)$  and  $\gamma_d(\tau)$  can be denoted by

$$\delta(\gamma_{y_k}, \gamma_d) := \int_{\alpha}^{\beta} \cdots \int_{\alpha}^{\beta} (\gamma_{y_k}(\tau) - \gamma_d(\tau))^2 d\tau, \quad (4)$$

the problem concerned in this paper can be formulated as follows.

The JPPDF Tracking Control Problem is to find control strategies  $u = g(x_k, r)$  such that

$$J_N := \sum_{k=0}^N \left[ Q_k \delta(\gamma_{y_k}, \gamma_d) + \frac{1}{2} u_k^T R_k u_k \right] \quad (5)$$

is minimized, and preferably, the closed loop system is stable, where  $Q_k > 0$  and  $R_k \geq 0$  are weighting matrices.

**Remark 1** The control target using the distance of JPPDFs is a new performance formulation existing in wide classes of engineering problems, where the desired output governs a known probabilistic distribution (see e. g. (Wang00)). It is noted that the measure for the distance in performance index  $J_N$  can be further generalized to  $L_p$ -distance, the Kullback-Leibler distance or entropy (Dev87, Pap91). Thus, the entropy optimization problem can also be dealt with similarly where  $\delta(\gamma_{y_k}, \gamma_d)$  can be changed to entropy of the tracking errors. Also, the proposed problem can study the classical tracking problem for non-Gaussian systems.

To simplify the controller design procedures, we introduce the following assumption.

**Assumption 4.** It is supposed that the Jacobian  $\Xi_k := \left| \det \frac{\partial H(x_k, u_k, w_k, v_k)}{\partial v_k} \right| \neq 0$ .

**Lemma 2** If Assumptions 1~4 hold, then we have

$$\gamma_{y_k}(\tau) = \int_a^b \gamma_v(H^{-1}(x_{1k}, u_k, \sigma, \tau)) \gamma_w(\sigma) \cdot \left| \det \frac{\partial H(x_{1k}, u_k, \sigma, \tau)}{\partial \tau} \right|^{-1} d\sigma$$

where  $H^{-1}(x_{1k}, u_k, \sigma, \tau)$  satisfies

$$\tau = H(x_{1k}, u_k, \sigma, H^{-1}(x_{1k}, u_k, \sigma, \tau)).$$

**Proof.** Based on Assumption 2 and Assumption 3,  $y_k$  ( $k = 0, 1, 2, \dots$ ) are also continuous random vectors, then the proof can be given by using the total probability lemma in (pap91).

## 3. RECURSIVE JPPDF TRACKING CONTROL STRATEGY

### 3.1 Design Based on Gradient Algorithms

In order to determine an optimal  $u_k$ , some optimization approaches can be applied to the JPPDF control problem. For simplicity of notations, we suppose  $Q_k = 1$  in (5) and only consider the explicit linear control strategies.

To provide recursive procedures, we denote

$$u_k = u_{k-1} + \Delta u_k, k = 1, 2, \dots, N, \dots, +\infty. \quad (6)$$

Based on (4) and (2.2), as a conditional JPPDF with respect to the previous states  $x_{ij}$ ,  $i = 1, 2$ ;  $j = 0, 1, 2, \dots, k - 1$ , disturbance inputs  $w_k$ ,  $v_k$  and control inputs  $u_k$ , the function  $\delta(\gamma_{y_k}, \gamma_d)$  can be approximated via

$$\delta(\gamma_{y_k}, \gamma_d) = \delta_0 + \delta_1 \Delta u_k + \Delta u_k^T \delta_2 \Delta u_k + o(\Delta u_k^2) \quad (7)$$

where  $\delta_0 := \delta(\gamma_{y_k}, \gamma_d)|_{u_k=u_{k-1}}$ ,  $\delta_1 := \frac{\partial \delta(\gamma_{y_k}, \gamma_d)}{\partial u_k} \Big|_{u_k=u_{k-1}}$ ,

$$\delta_2 := \frac{1}{2} \frac{\partial^2 \delta(\gamma_{y_k}, \gamma_d)}{\partial u_k^2} \Big|_{u_k=u_{k-1}}.$$

**Theorem 1.** Under Assumptions 1~4, the recursive suboptimal JPPDF control strategy to minimize the cost function  $J_N$  subject to descriptor system (3) is given by

$$\Delta u_k^* = -\frac{\delta_1 + R_k u_{k-1}}{2\delta_2 + 2R_k} \quad (8)$$

where  $R_k$  satisfies  $\delta_2 + R_k > 0$ .

**Proof.** The optimal control strategy can be obtained via

$$\frac{\partial \left[ \delta(\gamma_{y_k}, \gamma_d) + \frac{1}{2} u_k^T R_k u_k \right]}{\partial \Delta u_k} = 0. \quad (9)$$

Based on Bellman's principle of optimality, the resulting control law leads to the optimal one for the whole process.

From (6), it can be seen that

$$R_k u_k^2 = R_k u_{k-1}^2 + 2R_k u_{k-1} \Delta u_k + R_k \Delta u_k^2. \quad (10)$$

Substituting (7) and (10) into (9), we can obtain recursive suboptimal control law (8) for  $k = 1, 2, \dots, N, \dots, +\infty$ .

It is noted that the above algorithm only results from a necessary condition for optimization. To guarantee sufficiency, the following second-order derivative should be satisfied

$$\frac{\partial^2 [\delta(\gamma_{y_k}, \gamma_d) + \frac{1}{2}u_k^T R_k u_k]}{\partial \Delta u_k^2} = 2(\delta_2 + R_k) > 0 \quad (11)$$

which can be guaranteed if  $R_k$  is selected sufficient large.

### 3.2 Stabilization Control Strategy

It is noted that the proposed JPDP tracking control strategy resulting from the gradient approach leads to a nonlinear closed loop system. Generally stability analysis and synthesis are relatively difficult for stochastic nonlinear systems (see *e. g.* (Mao02, UP01, Wang02)), especially for multi-variable cases. In this section, an improved suboptimal optimization strategy will be proposed with which the closed loop system can be guaranteed to be stable.

For this purpose, under *Assumption 1*, we can also transform the state equation of system (1) to be (see also *e.g.*(LC99)

$$E x_{k+1} = \bar{A}_k x_k + \bar{B}_{1k} u_k + \bar{B}_{2k} w_k, \quad (12)$$

where  $x_k := [x_{1k}^T \ x_{2k}^T]^T$  and

$$E := \begin{bmatrix} I & 0 \\ 0 & 0 \end{bmatrix}, \quad \bar{A}_k := \begin{bmatrix} A_{1k} & 0 \\ 0 & -I \end{bmatrix},$$

$$\bar{B}_{1k} := \begin{bmatrix} B_{11k} \\ B_{12k} \end{bmatrix}, \quad \bar{B}_{2k} := \begin{bmatrix} B_{21k} \\ B_{22k} \end{bmatrix},$$

which leads to

$$E \Delta x_{k+1} = \bar{A}_k \Delta x_k + \bar{B}_{1k} \Delta u_k + \bar{B}_{2k} \Delta w_k, \quad (13)$$

where

$$\Delta x_k = x_k - x_{k-1}, \Delta u_k = u_k - u_{k-1}, \Delta w_k = w_k - w_{k-1}.$$

Equation (12) is also called Weierstrass form of  $\Sigma_D$ . To consider the stabilization problem, we introduce the following concept.

**Definition 2** If there exists a gain matrix  $C_k$  together with feedback law  $u_k = C_k x_{1k}$  such that  $E x_{k+1} = \bar{A}_k x_k + \bar{B}_{1k} u_k$  is stable, regular and impulse-free, the triple  $(E, \bar{A}_k, \bar{B}_{1k})$  denoted in (12) is called *generalized stabilizable* and  $u_k = C_k x_{1k}$  is called *generalized stabilization control law*.

The following assumption is necessary for stabilization problems, which corresponds to the stabilizable condition.

**Assumption 5.** The triple  $(E, \bar{A}_k, \bar{B}_{1k})$  in (12) is supposed to be *generalized stabilizable*.

**Lemma 3.** Under *Assumption 5*,  $u_k = C_k x_{1k}$  is the *generalized stabilization control law* of (12) if and only if there exists a symmetric matrix  $P$  and a gain matrix  $C_k$  satisfying

$$\tilde{A}_k^T P \tilde{A}_k < E^T P E \quad (14)$$

and  $E^T P E \geq 0$ , where

$$\tilde{A}_k := \bar{A}_k + \bar{B}_{1k} \bar{C}_k, \bar{C}_k := [C_k \ 0]$$

which is compatible to  $x_k$ .

**Proof.** It can be given by connecting Definition 1 with Theorem 2 of (HL99).

**Lemma 4.** For the transformed system (12) satisfying *Assumption 5*, at each sample time  $k$ , both of the following statements hold: (i) linear matrix inequality (LMI)

$$\begin{bmatrix} -Q & Q A_{1k}^T + S_1^T B_{11k}^T \\ A_{1k} Q + B_{11k} S_1 & -Q \end{bmatrix} < 0 \quad (15)$$

is solvable for definite positive matrix  $\dot{Q}$  and invertible matrix  $S_1$ . (ii) For given matrices  $\Gamma_{k1}$  and  $\Gamma_{k2}$ ,

$$\begin{bmatrix} -Q & P_k^T \\ P_k & -Q \end{bmatrix} < 0 \quad (16)$$

where  $P_k = A_{1k} Q + B_{11k} S_1 + B_{11k} S_2 \Gamma_{k2}$  is solvable for definite positive matrices  $\dot{Q}$  and  $S_2$ , and invertible matrix  $S_1$ . In both cases,  $\Delta u_k = S_1 Q^{-1} \Delta x_{1k}$  is the generalized stabilization control law.

**Proof.** See (HL99).

### 3.3 Stabilization JPDP Tracking Control Strategy

At this stage, the tracking control strategy can be constructed by

$$\Delta u_k = C_{1k} \Delta x_{1k} + \Delta u_k^* \quad (17)$$

where  $C_{1k}$  is the generalized stabilization control law given by *Lemma 4*, and  $\Delta u_k^*$  is the part for JPDP tracking control to be further determined.

Noting that  $\delta(\gamma_{y_k}, \gamma_d)$  is also a function of  $\Delta x_{1k}$ ,  $u_{k-1}$  and  $\Delta u_k$ , we consider the expansion

$$\delta(\gamma_{y_k}, \gamma_d) = \alpha_{k0} + \alpha_{k1} \Delta u_k + \alpha_{k2} \Delta x_{1k} + \frac{1}{2} \Delta u_k^T \Gamma_{k1} \Delta u_k$$

$$+ \Delta x_{1k}^T \Gamma_{k2} \Delta u_k + \frac{1}{2} \Delta x_{1k}^T \Gamma_{k3} \Delta x_{1k} + o(\Delta u_k^2, \Delta x_{1k}) \quad (18)$$

where

$$\begin{aligned} \alpha_{k0} &= \delta(\gamma_{y_k}, \gamma_d)|_{k-1}, \quad \Gamma_{k1} = \frac{\partial^2 \delta(\gamma_{y_k}, \gamma_d)}{\partial u_k^2} \Big|_{k-1}, \\ \alpha_{k1} &= \frac{\partial \delta(\gamma_{y_k}, \gamma_d)}{\partial u_k} \Big|_{k-1}, \quad \Gamma_{k2} = \frac{\partial^2 \delta(\gamma_{y_k}, \gamma_d)}{\partial x_{1k} \partial u_k} \Big|_{k-1}, \\ \alpha_{k2} &= \frac{\partial \delta(\gamma_{y_k}, \gamma_d)}{\partial x_{1k}} \Big|_{k-1}, \quad \Gamma_{k3} = \frac{\partial^2 \delta(\gamma_{y_k}, \gamma_d)}{\partial x_{1k}^2} \Big|_{k-1} \end{aligned} \quad (19)$$

are functions with respect to sample value  $\tau$ .

For flexibility of design, in the following,  $R_k$  will be determined at every sample step to guarantee stability of the closed loop systems.

*Theorem 2.* For the transformed system (12) satisfying *Assumptions 1 ~ 5*, at each sample time  $k$ , if we select

$$R_k = -S_2^{-1}I - \Gamma_{k1} \geq 0 \quad (20)$$

where  $S_2$  is calculated via (16), then the closed loop descriptor system is stable, and

$$\Delta u_k^* = -(\Gamma_{k1} + R_k)^{-1} (\Gamma_{k2}^T \Delta x_{1k} + \alpha_{k1}^T + R_k u_{k-1}) \quad (21)$$

together with (17) forms the stabilization JPDP tracking control strategy, where  $C_{1k}$  can be calculated by *Lemma 4*.

**Proof.** Equation(21) can be obtained by substituting (18) into (9) and removing the second-order terms. Substituting (21) into (13) yields the closed loop system described by

$$\begin{cases} \Delta x_{1k+1} = D_{1k} \Delta x_{1k} + B_{21k} \Delta w_k - p_{1k} \\ \Delta x_{2k} = D_{2k} \Delta x_{1k} + B_{22k} \Delta w_k - p_{2k} \end{cases} \quad (22)$$

where

$$\begin{aligned} D_{1k} &= A_{1k} + B_{11k} C_k - B_{11k} (\Gamma_{k1} + R_k)^{-1} \Gamma_{k2}^T, \\ D_{2k} &= A_{2k} + B_{21k} C_k - B_{21k} (\Gamma_{k1} + R_k)^{-1} \Gamma_{k2}^T \end{aligned}$$

and  $p_{ik} = B_{i1k} (\Gamma_{k1} + R_k)^{-1} (\alpha_{k1}^T + R_k u_{k-1})$ ,  $i = 1, 2$ , can be regarded as an additive bounded inputs. Similarly to proof of *Lemma 4*, it can be claimed that stability of descriptor system (22) depends on

$$\begin{aligned} &A_{1k} + B_{11k} C_k - B_{11k} (\Gamma_{k1} + R_k)^{-1} \Gamma_{k2}^T \\ &= A_{1k} + B_{11k} C_k + B_{11k} S_2 \Gamma_{k2}^T \end{aligned}$$

whose stability has been guaranteed based on *Lemma 4*.

**Remark 2.** The recursive suboptimal stabilization JPDP tracking control algorithms can be summarized as follows:

- Initialize  $x_0$  and  $u_0$ ;
- At the sample time  $k$ , compute  $C_k$  and  $S_{2k}$  by using (16);
- Formulate  $\gamma_{y_k}(\tau)$  using Lemma 2 and  $\delta(\gamma_{y_k}, \gamma_d)$  using (4);

- Calculate  $\Delta u_k^*$  and  $u_k$  using equation (21) and (6);
- Increase  $k$  by 1 to the next step.

**Remark 3.** In practice, generally nonlinearity also exists in many descriptor systems. In this paper, besides the above linear plants (1) or (3), the following nonlinear descriptor systems

$$\begin{cases} x_{1k+1} = f_{1k}(x_{1k}) + g_{11}(x_{1k})w_k + g_{21}(x_{1k})u_k \\ x_{2k} = f_{2k}(x_{1k}) + g_{21}(x_{1k})w_k + g_{22}(x_{1k})u_k \end{cases} \quad (23)$$

with a linear controlled output

$$y_k = H_k(x_{1k}, w_k, u_k, v_k, r) \quad (24)$$

can also be studied similarly, where  $x_k, x_{ik}$  ( $i = 1, 2$ ),  $w_k, u_k, v_k, y_k$  and  $r$  are defined as above.

#### 4. A NUMERAL EXAMPLE

In this section we give an example to demonstrate how to compute the JPDP tracking control laws. The considered descriptor system is given by

$$\begin{cases} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_{11k+1} \\ x_{12k+1} \\ x_{2k+1} \end{bmatrix} = \begin{bmatrix} 0.95 & 0 & 0 \\ 1 & -0.90 & 0 \\ 1 & -1 & -1 \end{bmatrix} \begin{bmatrix} x_{11k} \\ x_{12k} \\ x_{2k} \end{bmatrix} \\ + \begin{bmatrix} 0 \\ 1 \\ -1 \end{bmatrix} u_k + \begin{bmatrix} 0 \\ -0.50 \\ 0 \end{bmatrix} w_k, \\ y_k = \frac{1}{1 + x_{11k}^3} u_k + x_{12k} - 2x_{2k} + \sqrt{v_k}. \end{cases}$$

This system has been transferred to be a Weierstrass form. Random variables  $w_k$  and  $v_k$  ( $k = 0, 1, 2, \dots$ ) are assumed to be mutually independent with known PDFs. The asymmetric PDF of  $v_k$  is defined by

$$\gamma_v(x) = \begin{cases} \frac{-6(x - 0.25)}{\sqrt{x}} & x \in [0, 0.25], \\ 0 & \text{else.} \end{cases} \quad (25)$$

while the PDF of  $w_k$  is given by

$$\gamma_w(x) = \begin{cases} -48(x^2 - x + \frac{3}{16}) & x \in [0.25, 0.75], \\ 0 & \text{else.} \end{cases} \quad (26)$$

In simulations, the initial conditions are set to be  $u_0 = 0, y_0 = -0.4$ . For every sample time  $k$ , the sequence of the suboptimal control input is demonstrated in *Figure 1*. To illustrate the effectiveness of the proposed method, the practical output JPDPs under control at several different sample times as well as the tracked target JPDP are given in *Figure 2* for comparisons. *Figure 3* displays the outputs and the tracking errors where it can be seen that the tracking errors decrease to a small value in a short time. These simulation results show that satisfactory dynamical tracking performance can be obtained using the proposed approach.

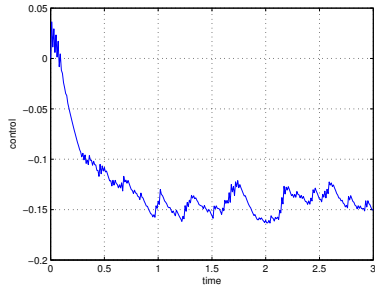


Fig. 1. JPDF control

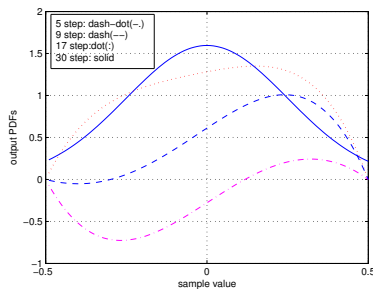


Fig. 2. the JPDF of the output

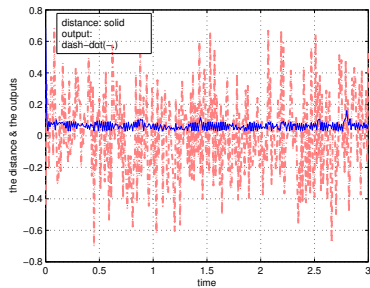


Fig. 3. the outputs and the tracking errors

## 5. CONCLUSION

In this paper, a new JPDF tracking control problem is considered for multivariate descriptor systems with non-Gaussian random inputs. Optimization approach is applied to present recursive algorithms such that the distances between the output distributions and the desired one are minimized. Furthermore, a stabilization suboptimal control strategy can be given by using of LMI-based Lyapunov theory. Simulation shows that the proposed approach can achieve satisfactory performance for non-Gaussian descriptor systems.

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