

Problem 1: Feedforward control

Consider the following process:

$$y = gu + g_d d \quad (1)$$

with a measured disturbance $d_m = g_{dm}d$. Propose a realizable feedforward controller when:

$$g(s) = 5 \frac{e^{-2s}}{5s+1} \quad (2)$$

$$g_d(s) = \frac{3}{5s+1} \quad (3)$$

$$g_{dm}(s) = \frac{1}{0.5s+1} \quad (4)$$

Inserting $d_m = g_{dm}d$

$$\Rightarrow d = \frac{d_m}{g_{dm}}$$

$$y = gu + g_d \cdot \frac{d_m}{g_{dm}}$$

An ideal feedforward controller can be found from $y=0$

$$\Rightarrow u = - \frac{g_d}{g \cdot g_{dm}} \cdot d_m$$

$$\Rightarrow C_{FF, ideal} = - \frac{g_d}{g \cdot g_{dm}} = - \frac{3}{5 \cdot \frac{e^{-2s}}{5s+1} \cdot \frac{1}{0.5s+1}} = - \frac{3}{5} (0.5s+1)e^{2s}$$

$$C_{FF}(s) = k \frac{(T_1s+1) \dots}{(\tau_1s+1)(\tau_2s+1) \dots} e^{-\theta s}$$

where must have at least as many τ 's as T 's

For the controller to be realizable, the pole-polynomial must have higher or equal order than the zero polynomial, also θ must be ≥ 0

selecting $(0.5s+1) \Rightarrow C_{FF} = \frac{C_{FF, ideal}(\theta=0)}{(0.5s+1)} = - \frac{3}{5} = -0.6$

Problem 2: Cascade control

The process below describes cascade control with controllers c_1 and c_2 based on measurements y_1 and y_2

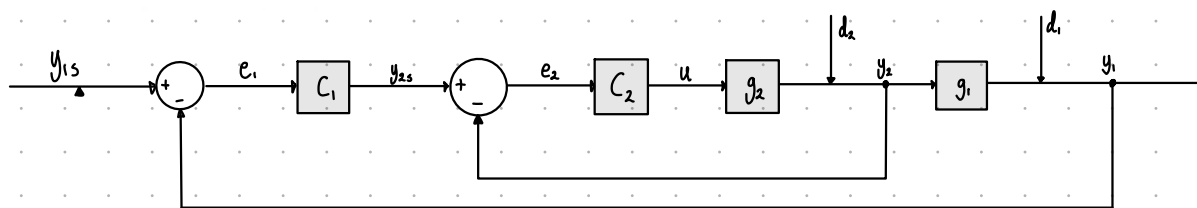
$$y_2 = g_2 u + d_2 \quad (5)$$

$$y_1 = g_1 y_2 + d_1 \quad (6)$$

$$u = c_2(y_2 - y_2) \quad (7)$$

$$y_{2s} = c_1(y_{1s} - y_1) \quad (8)$$

1. Draw a block diagram of cascade control.



2. Answer if the following statements are true or false. Justify your answers.

Cascade control is recommended when ...

- (a) d_1 is the main disturbance
- (b) d_2 is the main disturbance
- (c) g_1 has a large effective delay
- (d) g_2 has a large effective delay
- (e) g_1 is nonlinear
- (f) g_2 is nonlinear

a) False, then we don't have a benefit of controlling y_2 .

b) True, then the fast control will ensure that d_2 is neglected.

c) True, then we will have a fast inner loop, and a slower outer loop.

d) False, then the inner loop will have a lot of instability.

e) False } When g_2 is non-linear, then the outer loop will "see" g_2 as linear, as the controller
 f) True } is handling the non-linearity. This is not the case if g_1 is nonlinear.

3. Now, you are going to tune the controllers for the cascade control configuration.

In all cases, select τ_c = effective delay.

Consider the following transfer functions:

$$g_1(s) = \frac{2e^{-s}}{10s+1} \quad (9)$$

$$g_2(s) = \frac{3e^{-0.2s}}{15s+1} \quad (10)$$

(a) Tune a SIMC PI controller $c_2(s)$ for $g_2(s)$

$$K_c = \frac{1}{3} \cdot \frac{15}{2 \cdot 0.2} = \frac{5}{0.4} = 12.5$$

$$\tau_I = \min(15, 4 \cdot (2 \cdot 0.2)) = 1.6$$

$$\Rightarrow \underline{\underline{c_2(s) = 12.5 \cdot \frac{1.6s+1}{1.6s}}}$$

■ First-order + delay model for PI-control

$$G(s) = \frac{k}{\tau_1 s + 1} e^{-\theta s}$$

■ PI-controller (based on first-order model)

$$c(s) = K_c \left(1 + \frac{1}{\tau_I s}\right) = K_c \frac{\tau_I s + 1}{\tau_I s}$$

For cascade form PID controller:

$$K_c = \frac{1}{k} \frac{\tau_1}{\tau_c + \theta} = \frac{1}{k'} \cdot \frac{1}{\tau_c + \theta}$$

$$\tau_I = \min\left\{\tau_1, \frac{4}{k' K_c}\right\} = \min\{\tau_1, 4(\tau_c + \theta)\}$$

(b) Obtain the closed loop transfer function $T_2(s)$ for the inner loop.

Hint: the closed loop response of the inner loop $T_2(s)$ can be approximated to a first order with time delay process. This can be used to tune controller $c_1(s)$ in part (c) and (d),

$$T_2(s) = \frac{c_2(s)g_2(s)}{1 + c_2(s)g_2(s)} \approx \frac{e^{-\theta s}}{\tau_c s + 1} \quad (11)$$

$$\text{Using eq. (11): } T_2(s) \approx \frac{e^{-0.2s}}{0.2s + 1}$$

(c) Obtain the transfer function $g'_1(s)$ for the "new" process from y_2 to y_1 .

$$g'_1(s) = T_2 \cdot g_1 = \frac{e^{-0.2s}}{0.2s + 1} \cdot \frac{2e^{-s}}{10s + 1}$$

$$\underline{\underline{g'_1(s) = 2 \cdot \frac{e^{-1.2s}}{(10s+1)(0.2s+1)}}}$$

(d) Tune a SIMC PI controller $c_1(s)$ for $g'_1(s)$.

$$\text{Using the half rule approximation: } \Theta = \theta_0 + \frac{\tau_2}{2} = 1.2 + 0.1 = 1.3$$

$$\tau_1 = 10 + \frac{0.2}{2} = 10.1$$

$$\Rightarrow g'_1(s) \approx 2 \cdot \frac{e^{-1.3s}}{10.1s + 1}$$

Then, the SIMC-rules gives ($\tau_c = \theta$)

$$K_c = \frac{1}{2} \cdot \frac{10,1}{2 \cdot 1,3} = 1,94$$

$$\tau_I = \min(10,1, 8 \cdot 1,3) = 10,1$$

Then,
$$C_1 = 1,94 \cdot \frac{10,1s + 1}{10,1s}$$

(e) What would the SIMC PI controller be **without** the cascade? In other words, tune a controller $c_3(s)$ for $g_3(s) = g_1(s)g_2(s)$.

$$g_3 = \frac{2e^{-s}}{10s + 1} \cdot \frac{3e^{-0,2s}}{15s + 1} = \frac{6e^{-1,2s}}{(15s+1)(10s+1)}$$

Using the half rule:

$$\tau_1 = 15 + \frac{10}{2} = 20, \quad \theta = 1,2 + \frac{10}{2} = 6,2$$

$$g_3 \approx 6 \cdot \frac{e^{-6,2s}}{20s + 1}$$

Then, using SIMC-rules:

$$K_c = \frac{1}{6} \cdot \frac{20}{2 \cdot 6,2} = 0,269$$

$$\tau_I = \min(20 + 4 \cdot (2 \cdot 6,2)) = 20$$

Problem 3: Water mixer

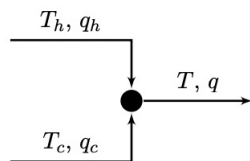


Figure 1: Mixer system

Consider the process of mixing hot and cold water, as shown in Figure 1. The process has inputs $u_1 = \Delta q_h$ [ℓ/s], $u_2 = \Delta q_c$ [ℓ/s], and outputs $y_1 = \Delta T$ [$^{\circ}C$], $y_2 = \Delta q$ [ℓ/s].

The control objective is to have a mixing temperature $T = 40^{\circ}C$ and a total flow leaving the mixer of $q = 1 \ell/s$. At the nominal operating point we have $T_c = 30^{\circ}C$ and $T_h = 60^{\circ}C$.

1. Formulate the energy and mass balances. The dynamics of this process are very fast; so, a steady-state model is sufficient to get T and q .

Mass balance: $q = q_h + q_c$

Energy balance: $q C_{p,w} T = q_c C_{p,w} T_c + q_h C_{p,w} T_h$ Assume $C_{p,w} = \text{const}$

$$T = \frac{q_c T_c + q_h T_h}{q} = \frac{q_c T_c + q_h T_h}{q_c + q_h}$$

2. Linearize the model and show that the linear model can be written $y = Gu$, where:

$$G = \begin{bmatrix} k_1 & k_2 \\ 1 & 1 \end{bmatrix}$$

with: $k_1 = (T_h^* - T^*)/q^*$ $k_2 = (T_c^* - T^*)/q^*$ $u = [u_1 \quad u_2]^T$

\Rightarrow Want $G = \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix}$

The symbol * denotes the steady state value.

Introducing deviation variables: $\Delta \Psi = \Psi - \Psi^*$

$$\Delta q = \Delta q_c + \Delta q_h \Rightarrow g_{21} = g_{22} = 1$$

$$\Delta T = g_{22} \Delta q_c + g_{21} \Delta q_h$$

Using the 1st order Taylor expansion:

$$g_{12} = \frac{\partial T}{\partial q_c} = \frac{T_c \cdot (q_c + q_h) - (q_c T_c + q_h T_h) \cdot 1}{(q_c + q_h)^2} = \frac{T_c}{q_c + q_h} - \frac{q_c T_c + q_h T_h}{(q_c + q_h)^2} = \frac{T_c - T}{q_c + q_h} = k_2$$

I expect a similar result for g_{21} due to how the expression looks:

$$g_{11} = \frac{T_h - T}{q_c + q_h} = k_1$$

In conclusion:

$$\Delta q = 1 \cdot \Delta q_c + 1 \cdot \Delta q_h$$

$$\Delta T = \underbrace{\frac{T_c - T}{q_c + q_h}}_{k_2} \Delta q_c + \underbrace{\frac{T_h - T}{q_c + q_h}}_{k_1} \Delta q_h$$

Can be written as $y = Gu$,
where $y = \begin{bmatrix} \Delta T \\ \Delta q \end{bmatrix}$, $u = \begin{bmatrix} \Delta q_h \\ \Delta q_c \end{bmatrix}$

$$G = \begin{bmatrix} k_1 & k_2 \\ 1 & 1 \end{bmatrix} \quad \underline{\underline{QED}}$$

3. What are the steady state values for q_c and q_h ?

$$(1) \quad q^* = 1 = q_c^* + q_h^* \Rightarrow q_c^* = 1 - q_h^*$$

$$(2) \quad T^* = 40 = \frac{q_c^* \cdot T_c^* + q_h^* \cdot T_h^*}{q^*} = \frac{(1 - q_h^*) \cdot 30 + q_h^* \cdot 60}{1}$$

$$40 = 30 + 30q_h^* \Rightarrow q_h^* = \frac{1}{3} \xrightarrow{(1)} q_c^* = \frac{2}{3}$$

4. Find the gain matrix G at the nominal operating point.

$$k_1^* = \frac{60 - 40}{1} = 20 \quad k_2^* = \frac{30 - 40}{1} = -10$$

$$G^* = \begin{bmatrix} 20 & -10 \\ 1 & 1 \end{bmatrix}$$

5. Based on G , which stream (q_h or q_c) would you use to control the temperature (T)? Explain briefly.

I would use q_h , as it has the highest gain due to the larger temperature difference, which gives a better/larger/faster response than for q_c .