

Simple PID tuning rules with guaranteed M_s robustness achievement

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Abstract: The design of the closed-loop control system must take into account the system performance to load-disturbance and set-point changes and its robustness to variation of the controlled process characteristics, preserving the well-known trade-off among all these variables. This paper face with this combined servo/regulation performance and robustness problem. The proposed method is formulated as an optimization problem for combined performance (not for independent operation modes), including also the robustness property as a constraint. The accomplishment of the claimed robustness is checked and then, the PID controller gives a good performance with also a precise and specific robustness degree. The proposed robust based PID control design is tested against other tuning methods.

Keywords: PID control, Process control, robustness, performance

1. INTRODUCTION

Since their introduction in 1940 (Babb, 1990; Bennett, 2000) commercial *Proportional - Integrative - Derivative* (PID) controllers have been with no doubt the most extensive option that can be found on industrial control applications (Åström and Hägglund, 2001). Their success is mainly due to its simple structure and to the physical meaning of the corresponding three parameters (therefore making manual tuning possible). This fact makes PID control easier to understand by the control engineers than other most advanced control techniques. In addition, the PID controller provides satisfactory performance in a wide range of practical situations.

With regard to the design and tuning of PID controllers, there are many methods that can be found in the literature over the last sixty years. Special attention is made of the *IFAC workshop PID'00 - Past, Present and Future of PID Control*, held in Terrassa, Spain, in April 2000, where a glimpse of the state-of-the-art on PID control was provided. Moreover, because of the widespread use of PID controllers, it is interesting to have simple but efficient methods for tuning the controller.

In fact, since the initial work of Ziegler and Nichols (1942), an intensive research has been done, developing autotuning methods to determine the PID controller parameters (Skogestad, 2003; Åström and Hägglund, 2004; Kristiansson and Lennartson, 2006). It can be seen that most of them are concerned with feedback controllers which are tuned either with a view to the rejection of disturbances (Cohen and Coon, 1953; López et al., 1967) or for a well-damped fast response to a step change in the controller set-point (Rovira et al., 1969; Martin et al., 1975; Rivera et al.,

1986).

Moreover, in some cases the methods considered only the system performance (Ho et al., 1999), or its robustness (Åström and Hägglund, 1984; Ho et al., 1995; Fung et al., 1998). However, the most interesting cases are the ones that combine performance and robustness, because they face with all system's requirements Ho et al. (1999); Ingimundarson et al. ((n.d.); Yaniv and Nagurka (2004); Vilanova (2008).

O'Dwyer (2003) presents a complete collection of tuning rules for PID controllers, which show their abundance.

Taking into account that in industrial process control applications, it is required a good load-disturbance rejection (usually known as *regulatory-control*), as well as, a good transient response to set-point changes (known as *servo-control* operation), the controller design should consider both possibilities of operation.

Despite the above, the servo and regulation demands cannot be simultaneously satisfied with a One-Degree-of-Freedom (1-DoF) controller, because the resulting dynamic for each operation mode is different and it is possible to choose just one for an optimal solution.

Considering the previous statement, the studies have focused only in fulfilling one of the two requirements, providing tuning methods that are optimal to servo-control or to regulation-control. However, it is well known that if we optimize the closed-loop transfer function for a step-response specification, the performance with respect to load-disturbance attenuation can be very poor and vice-versa (Arrieta and Vilanova, 2010). Therefore, it is desirable to get a compromise design, between servo/regulation, by using 1-DoF controller.

The proposed method considers 1-DoF PID controllers as

an alternative when *explicit* Two-Degree-of-Freedom (2-DoF) PID controllers are not available. Therefore, it could be stated that the proposed tuning can be used when both operation modes may happen and it could be seen as an *implicit* 2-DoF approach (because the design takes into account both objectives, servo and regulation modes) (Arrieta et al., 2010).

Moreover, it is important that every control system provides a certain degree of robustness, in order to preserve the closed-loop dynamics, to possible variations in the process. Therefore, the robustness issue should be included within the multiple *trade-offs* presented in the control design and it must be solved on a balanced way.

With respect to the robustness issue, during the last years, there has been a perspective change of how to include the robustness considerations. In this sense, there is variation from the classical Gain and Phase Margin measures to a single and more general quantification of robustness, such as the Maximum of the Sensitivity function magnitude.

Taking also into account the importance of the explicit inclusion of robustness into the design, the aim is to look for an optimal tuning for a combined servo/regulation index, that also guarantees a robustness value, specified as a desirable Maximum Sensitivity requirement.

The previous cited methods study the performance and robustness jointly in the control design. However, no one treats specifically the performance/robustness *trade-off* problem, nor consider in the formulation the servo/regulation *trade-off* or the interacting between all of these variables.

In addition, this work raises a point not found in the literature, such as the fulfillment of the robustness specification. Many of the existing tuning rules that include robustness constraints never check the accomplishment of such constraint. Therefore, the robustness of the resulting control system is not known. On this respect, we attempt to generate tuning rules (in fact simple tuning rules) that fulfill such constraint, providing at the same time, the maximum combined servo/regulation performance. Therefore, it can be stated the main contribution presented in this paper.

The paper is organized as follows. Section 2 introduces the control system configuration, as well as the general framework. The optimization problem setup is stated in Section 3, where it is defined the combined servo/regulation performance index and the robustness constraint. The proposed PID tuning with guaranteed robustness achievement is in Section 4 and some examples are provided in Section 5. The paper ends in Section 6 with some conclusions.

2. GENERAL FRAMEWORK

2.1 Control system configuration

We consider the feedback control system shown in Fig. 1, where $P(s)$ is the controlled process, $C(s)$ is the controller, $r(s)$ is the set-point, $u(s)$ is the controller output signal, $d(s)$ is the load-disturbance and $y(s)$ is the system output. The output of the ISA-PID controller Åström and Hägglund (2006) is given by

$$u(s) = K_p \left(1 + \frac{1}{T_i s} \right) e(s) - K_p \left(\frac{T_d s}{1 + (T_d/N)s} \right) y(s) \quad (1)$$

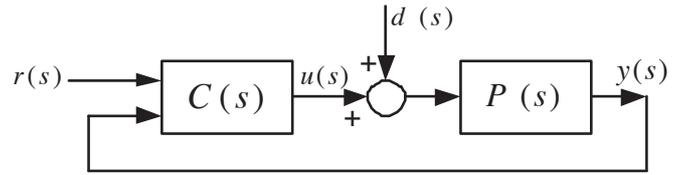


Figure 1. Closed-loop control system.

where $e(s) = r(s) - y(s)$ is the control error, K_p is the controller static gain, T_i the integral time constant, T_d the derivative time constant and the derivative filter constant N is taken $N = 10$ as it is usual practice in industrial controllers.

Also, the process $P(s)$ is assumed to be modelled by a First-Order-Plus-Dead-Time (FOPDT) transfer function of the form

$$P(s) = \frac{K}{1 + Ts} e^{-Ls} \quad (2)$$

where K is the process gain, T is the time constant and L is the dead-time. This model is commonly used in process control because is simple and describes the dynamics of many industrial processes approximately (Åström and Hägglund, 2006).

The availability of FOPDT models in the process industry is a well known fact. The generation of such model just needs for a very simple step-test experiment to be applied to the process. From this point of view, to maintain the need for plant experimentation to a minimum is a key point when considering industrial application of a technique.

2.2 Performance

One way to evaluate the performance of control systems is by calculating a cost function based on the error, i.e. the difference between the desired value (set-point) and the actual value of the controlled variable (system's output). Of course, as larger and longer in time is the error, the system's performance will be worse.

In this sense, a common reference for the evaluation of the controller performance, is a functional based on the integral of the error like: Integral-Square-Error (ISE), or Integral-Absolute-Error (IAE).

Some approaches had used the ISE criterion, because its definition allows an analytical calculation for the index (Zhuang and Atherton, 1993). However, nowadays can be found in the literature that IAE is the most useful and suitable index to quantify the performance of the system (Chen and Seborg, 2002; Skogestad, 2003; Åström and Hägglund, 2006; Kristiansson and Lennartson, 2006; Tan et al., 2006). It can be used explicitly in the design stage or just as an evaluation measure.

The formulation of the criterion is stated as

$$IAE \doteq \int_0^{\infty} |e(t)| dt = \int_0^{\infty} |r(t) - y(t)| dt \quad (3)$$

where the index can be measure for changes in the set-point or in the load-disturbance.

2.3 Robustness

Robustness is an important attribute for control systems, because the design procedures are usually based on the

use of low-order linear models identified at the closed-loop operation point. Due to the non-linearity found in most of the industrial process, it is necessary to consider the expected changes in the process characteristics assuming certain relative stability margins, or robustness requirements, for the control system.

As an indication of the system *robustness* (relative stability) the Sensitivity Function peak value will be used. The control system Maximum Sensitivity is defined as

$$M_s \doteq \max_{\omega} |S(j\omega)| = \max_{\omega} \frac{1}{|1 + C(j\omega)P(j\omega)|} \quad (4)$$

and recommended values for M_s are typically within the range 1.4 - 2.0 Åström and Hägglund (2006). The use of the maximum sensitivity as a robustness measure, has the advantage that lower bounds to the Gain, A_m , and Phase, ϕ_m , margins (Åström and Hägglund, 2006) can be assured according to

$$A_m > \frac{M_s}{M_s - 1} \quad ; \quad \phi_m > 2 \sin^{-1} \left(\frac{1}{2M_s} \right)$$

Therefore, to assure $M_s = 2.0$ provides what is commonly considered minimum robustness requirement (that translates to $A_m > 2$ and $\phi_m > 29^\circ$, for $M_s = 1.4$ we have $A_m > 3.5$ and $\phi_m > 41^\circ$).

In many cases, robustness is specified as a target value of M_s , however the accomplishment of the resulting value is never checked.

3. OPTIMIZATION PROBLEM SETUP

From the above definitions for performance and robustness specifications, there appears the need to formulate a joint criteria that faces with the *trade-off* between the performance for servo and regulation operation and also that takes into account the accomplishment of a robustness level.

3.1 Servo/Regulation trade-off

As it is known, there is a *trade-off* behavior between the dynamics for servo and regulation control operation modes. It is not enough just to consider the tuning mode, it is also necessary to include the system operation in the controller's design.

Using some of the exposed ideas, we can say that J_x^z represents the criteria (3) taking into account the operation mode x , for a tuning mode z . From this, we can post the following definitions:

- J_r^r is the value of performance index for the set-point tuning operating in servo-control mode.
- J_d^r is the value of performance index for the set-point tuning operating in regulatory-control mode.
- J_r^d is the value of performance index for the load-disturbance tuning operating in servo-control mode.
- J_d^d is the value of performance index for the load-disturbance tuning operating in regulatory-control mode.

Obviously J_r^r is the optimal value for servo-control operation, J_r^o , and J_d^d is the optimal one for regulation, J_d^o . An *intermediate* tuning between servo and regulation operation modes should have higher values than the optimal ones, when the tuning and operation modes are the

same, but the indexes would be lower when the modes are different. So, for each operation mode we have the following relationships,

$$J_r^o \doteq J_r^r \leq J_r^{rd} \leq J_r^d$$

$$J_d^o \doteq J_d^d \leq J_d^{rd} \leq J_d^r$$

where J_r^{rd} and J_d^{rd} are the performance values of the *intermediate* tuning for servo and regulation control operation, respectively.

The previous ideas can be represented graphically, the results are shown in Fig. 2, where the performance indexes are plotted in the $J_r - J_d$ plane. It can be seen that the

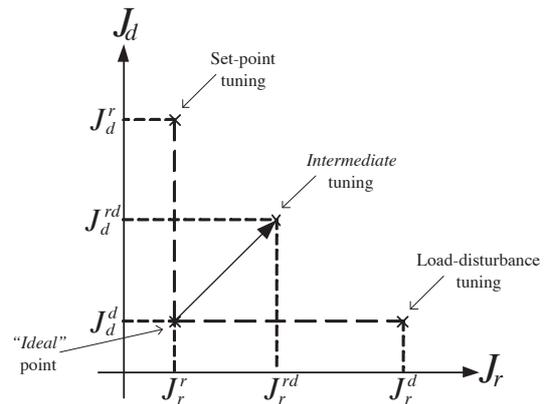


Figure 2. Plane $J_r - J_d$.

point (J_r^r, J_d^d) is the “ideal” one because it represents the minimum performance values taking both possible operation modes, servo and regulation, into account. However, this point is unreachable due the differences in the dynamics for each one of the objectives of the control operation modes. Therefore our efforts must go towards getting the minimum resulting distance, meaning the best balance between the operation modes.

On this way, a cost objective function is formulated in order to get closer, as much as possible, the resulting point (J_r^{rd}, J_d^{rd}) , to the “ideal” one, (J_r^o, J_d^o) . Therefore,

$$J_{rd} = \sqrt{(J_r^{rd} - J_r^o)^2 + (J_d^{rd} - J_d^o)^2} \quad (5)$$

where J_r^o and J_d^o are the optimal values for servo and regulation control respectively, and J_r^{rd} , J_d^{rd} are the performance indexes for the *intermediate* tuning considering both operation modes. In Fig. 2, index (5) is represented by the arrow between the “ideal” point and the corresponding to the *intermediate* tuning.

From the above analysis, the optimization problem setup considers the model's normalized dead-times, τ , in the range $0.1 \leq \tau \leq 2.0$, to obtain the PID controller optimum parameters such that

$$\bar{p}_o := [K_{po}, T_{io}, T_{do}] = \arg \left[\min_{\bar{p}} J_{rd} \right] \quad (6)$$

where \bar{p} is the PID controller parameters vector. Here, optimization is done using genetic algorithms technique (Mitchell, 1998).

The aim of minimizing (6) is to achieve a balanced performance for both operation modes of the control system.

3.2 Robustness constraint criterion

The cost functional (5) proposed before, even though face with the *trade-off* problem between the operation modes of the system, just takes into account characteristics of performance. However, there is a need to include a certain robustness for the control-loop.

In that sense, we want to use (4) as a robustness measure. So, the optimization problem (6) is subject to a constraint of the form

$$|M_s - M_s^d| = 0 \quad (7)$$

where M_s and M_s^d are the Maximum Sensitivity and the desired Maximum Sensitivity functions respectively. This constraint tries to guarantee the selected robustness value for the control system.

4. SERVO/REGULATION PID TUNINGS WITH ROBUSTNESS CONSIDERATION

From the previous formulation, we look for a tuning rule that faces with the *trade-off* problem between the performance for servo and regulation modes and providing, at the same time, a certain degree of robustness (if necessary).

As it has been stated, we solve the optimization problem (6) subject to constraint (7). In that sense, a broad classification can be established, using specific values for M_s , within the suggested range between 1.4 – 2.0. This will allow a qualitative specification for the control system robustness. So, the rating is described here as

- Low robustness level - $M_s = 2.0$
- Medium-low robustness level - $M_s = 1.8$
- Medium-high robustness level - $M_s = 1.6$
- High robustness level - $M_s = 1.4$

According to this principle, the above mentioned four values for M_s are used here as desirable robustness, M_s^d into the robustness constraint (7), for the problem optimization (6). Additionally, an unconstrained optimization is done, that can be seen as the M_s^d free case.

In order to provide results for autotuning methodology, the optimal sets for the PID parameters with the corresponding desired robustness, are approximated in equations for each controller's parameter. This fitting procedure looks for simple expressions that allow for an homogenized set, to preserve the simplicity and completeness of the approach.

Therefore, the resulting controller parameters will be, expressed just in terms of the FOPDT process model parameters (2) as

$$\begin{aligned} K_p K &= a_1 \tau^{b_1} + c_1 \\ \frac{T_i}{T} &= a_2 \tau^{b_2} + c_2 \\ \frac{T_d}{T} &= a_3 \tau^{b_3} + c_3 \end{aligned} \quad (8)$$

where the constants a_i , b_i and c_i are given in Table 1, according to the desired robustness level for the control system.

It is important to note that, although there may be other tuning equations that provide a good fit, the choice of the proposals (8) represents an option to retain the simplicity

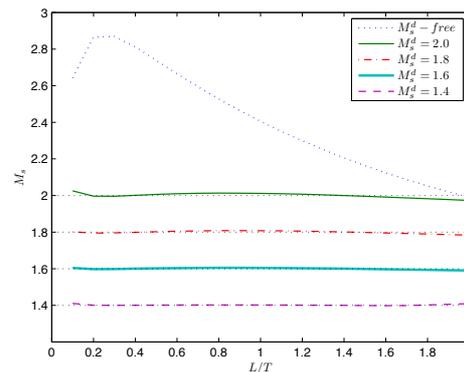


Figure 3. Accomplishment for each claimed robustness level.

that is searched, because all the robustness levels and controller's parameters can be expressed in the same form and only changing the constants, according to each case as in Table 1.

Table 1. PID settings for servo/regulation tuning with robustness consideration

| constant | M_s^d free | $M_s^d = 2.0$ | $M_s^d = 1.8$ | $M_s^d = 1.6$ | $M_s^d = 1.4$ |
|----------|--------------|---------------|---------------|---------------|---------------|
| a_1 | 1.1410 | 0.7699 | 0.6825 | 0.5678 | 0.4306 |
| b_1 | -0.9664 | -1.0270 | -1.0240 | -1.0250 | -1.0190 |
| c_1 | 0.1468 | 0.3490 | 0.3026 | 0.2601 | 0.1926 |
| a_2 | 1.0860 | 0.7402 | 0.7821 | 0.8323 | 0.7894 |
| b_2 | 0.4896 | 0.7309 | 0.6490 | 0.5382 | 0.4286 |
| c_2 | 0.2775 | 0.5307 | 0.4511 | 0.3507 | 0.2557 |
| a_3 | 0.3726 | 0.2750 | 0.2938 | 0.3111 | 0.3599 |
| b_3 | 0.7098 | 0.9478 | 0.7956 | 0.8894 | 0.9592 |
| c_3 | -0.0409 | 0.0034 | -0.0188 | -0.0118 | -0.0127 |

In the literature, there are many control designs that include robustness in the formulation stage and even more, in some cases the consideration is regarded as a parameter design directly. However, none of these methods check the accomplishment of the claimed robustness and this should be an aspect that deserves much attention.

The deviation of the resulting value of M_s with respect to the specified target has a direct influence (as a *trade-off*) in the performance of the system (Vilanova et al., 2010).

In order to guarantee the selected robustness, the constraint stated in (7) forces the optimization problem to fulfill the fixed value M_s^d and for this, the minimum of the performance index J_{rd} is achieved.

Here, the resulting robustness, applying the proposed methodology, is compared to the desired one, in order to check the accomplishment of the claimed robustness. Fig. 3 shows that the robust tuning has a very good accuracy for the M_s values for all the range of processes, therefore assuring that performance is the best one that can be achieved for that robustness value. From the very well known performance-robustness *trade-off*, the increase of the system's robustness from the M_s^d -free case (no robustness constraint), is reflected in a deterioration of the system's performance, and vice-versa. Similar to Fig. 3, where it can be seen the robustness increasing, in Fig. 4 it is shown how the performance index J_{rd} varies, for each one of the proposed robustness levels. If we use the information of Fig. 3 and Fig. 4, and the unconstrained case as the starting point, it is possible to see that for

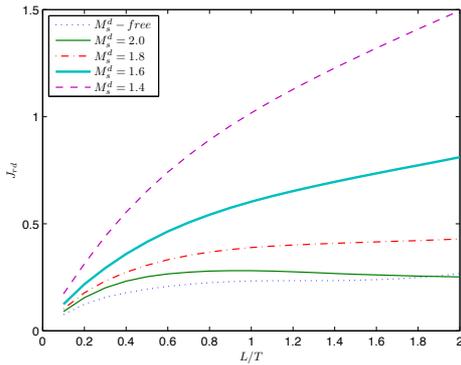


Figure 4. Combined index J_{rd} for each robustness level tuning.

each selected level, the robustness is improved achieving smaller values for M_s , but at the same time having larger values (i.e worse) for the performance index J_{rd} .

It is also important to note that, the relation between the loss of performance and the robustness increase (for each level of M_s) is nonlinear, neither for the τ range. For example, in Fig. 4 the difference between the performance for cases $M_s^d = 1.8$ and $M_s^d = 1.6$, is much smaller than the one for $M_s^d = 1.6$ and $M_s^d = 1.4$, despite that the levels are equally separated.

In general terms, it is possible to say that the robustness requirements are fulfilled, facing at the same time, to the performance servo/regulation *trade-off* problem.

5. COMPARATIVE EXAMPLE

This section presents an example in order to evaluate the characteristics of the proposed tuning rule. The analysis is not only for a specific process, but for the whole set of plants provided in their range of validity, $\tau \in [0.1, 2.0]$, in order to show the global advantages that the proposal can provide.

The robust tuning rules that can be found in the literature consider different specifications for M_s . They range from the considered minimum robustness; $M_s = 2.0$; to a high robustness; $M_s = 1.4$.

Here, we compare the tuning proposed in Section 4 with the following methods:

- AMIGO method (Åström and Hägglund, 2004) provides tuning with a design specification of $M_s = 1.4$.
- Kappa-Tau ($\kappa - \tau$) method (Åström and Hägglund, 1995) provides tuning with a design specification of $M_s = 1.4$ and $M_s = 2.0$.
- Tavakoli method (Tavakoli et al., 2005) provides tuning with a design specification of $M_s = 2.0$.

Fig. 5 shows the achieved M_s values for 1.4 and 2.0 cases, for the compared tuning rules. With this information and the one in Fig. 3, it seems that the proposed tuning is the option that provides the best accuracy for the selected robustness. With the aim to establish a more precise and quantitative measure of the claimed robustness accomplishment for the whole range of models, the next index is stated

$$I_{M_s} \doteq \int_{\tau_o}^{\tau_f} |M_s^r(\tau) - M_s^d| d\tau \quad (9)$$

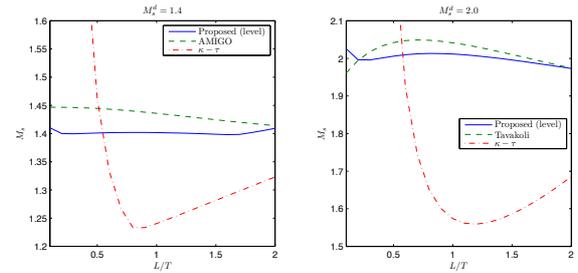


Figure 5. Comparative for claimed robustness.

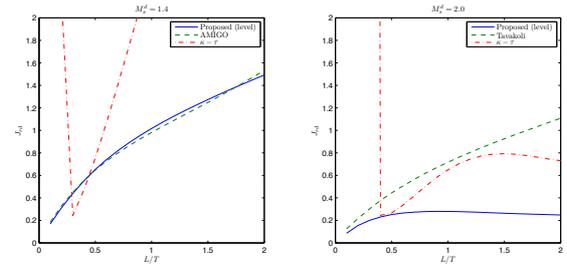


Figure 6. J_{rd} index for the compared tuning rules.

where M_s^d and M_s^r are the desired and resulting M_s values, respectively. As I_{M_s} is smaller, the accuracy is better. In Table 2 there are the values (9) for the analyzed tuning rules. Now, from the plots in Figs. 3 and 5, and the measured values (9) in Table 2, it is possible to say that the proposed PID tuning (using the levels classification), is the one that provides the best accomplishment between the desirable and the achieved robustness.

Once the robustness accomplishment has been verified, it is important to see the resulting performance for the compared tuning rules. Fig. 6 shows the combined servo/regulation performance index (5). For the $M_s^d = 2.0$ it is obvious that the proposed tuning is the one that provides the best robustness accomplishment and at the same time, the best achievable performance. For $M_s^d = 1.4$ case, because AMIGO method does not fulfill the robustness requirements, having a somewhat lower robustness, it has values slightly lower for J_{rd} index compared to the proposed tuning. However, the proposal is more accurate for the claimed robustness with also good performance (see Fig. 6). This fact strongly confirms the importance of the relation between robustness and performance variations.

6. CONCLUSIONS

In process control, it is very important to guarantee some degree of robustness, in order to preserve the closed-loop

Table 2. Claimed robustness accomplishment I_{M_s} for different tuning rules

| Tuning | M_s^d | I_{M_s} |
|-------------------|---------|-----------|
| AMIGO | 1.4 | 0.0634 |
| $\kappa - \tau$ | 1.4 | 4.9672 |
| $\kappa - \tau$ | 2.0 | 10.2377 |
| Tavakoli | 2.0 | 0.0519 |
| Proposed (levels) | 1.4 | 0.0035 |
| | 1.6 | 0.0066 |
| | 1.8 | 0.0111 |
| | 2.0 | 0.0186 |

dynamics, to possible variations in the control system. Also, at the same time, it must be provided the best achievable performance for servo and regulation operation. All of the above specifications, lead to have different *trade-offs*, between performance and robustness or between servo and regulation modes, that must be solved on a balanced way. Here, we looked for a PID controller tuning rule that faces to the general problem. This tuning is optimal, as much as possible, to a proposed performance index that takes into account both system operation modes, including also a certain degree of robustness, specified as a desirable Maximum Sensitivity value.

Autotuning formulae have been presented for two approaches. First, robustness is established using a qualitative levels classification and then, the idea is extended to an issue that offers a generic expression, to allow the specification in terms of any value of robustness in the range $M_s \in [1.4, 2.0]$. Moreover, taking into account the performance/robustness *trade-off*, the accuracy of the claimed robustness is a point that has been verified, achieving flat curves for the resulting values. In short, both approaches are the main contributions presented in this paper.

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REFERENCES

- Arrieta, O. and Vilanova, R. (2010). Performance degradation analysis of controller tuning modes: Application to an optimal PID tuning. *International Journal of Innovative Computing, Information and Control*, 6(10), 4719–4729.
- Arrieta, O., Visioli, A., and Vilanova, R. (2010). PID autotuning for weighted servo/regulation control operation. *Journal of Process Control*, 20(4), 472–480.
- Åström, K. and Hägglund, T. (1984). Automatic tuning of simple regulators with with specifications on phase and amplitude margin. *Automatica*, 20, 645–651.
- Åström, K. and Hägglund, T. (1995). *PID Controllers: Theory, Design, and Tuning*. Instrument of Society of America.
- Åström, K. and Hägglund, T. (2001). The future of PID control. *Control Engineering Practice*, 9, 1163–1175.
- Åström, K. and Hägglund, T. (2004). Revisiting the Ziegler-Nichols step response method for PID control. *Journal of Process Control*, 14, 635–650.
- Åström, K. and Hägglund, T. (2006). *Advanced PID Control*. ISA - The Instrumentation, Systems, and Automation Society.
- Babb, M. (1990). Pneumatic Instruments Gave Birth to Automatic Control. *Control Engineering*, 37(12), 20–22.
- Bennett, S. (2000). The Past of PID Controllers. In *IFAC Digital Control: Past, Present and Future of PID Control*. Terrassa, Spain.
- Chen, D. and Seborg, D. (2002). PI/PID controller design based on direct synthesis and disturbance rejection. *Industrial Engineering & Chemistry Research*, 41(19), 4807–4822.
- Cohen, G.H. and Coon, G.A. (1953). Theoretical Considerations of Retarded Control. *ASME Transactions*, 75, 827–834.
- Fung, H., Wang, Q., and Lee, T. (1998). PI Tuning in terms of gain and phase margins. *Automatica*, 34, 1145–1149.
- Ho, W.K., Hang, C.C., and Cao, L.S. (1995). Tuning PID Controllers Based on Gain and Phase Margin Specifications. *Automatica*, 31(3), 497–502.
- Ho, W.K., Lim, K.L., Hang, C.C., and Ni, L.Y. (1999). Getting more phase margin and performance out of PID controllers. *Automatica*, 35, 1579–1585.
- Ingimundarson, A., Hägglund, T., and Åström, K.J. ((n.d.)). Criteria for desing of PID controllers. Technical report, ESAII, Universitat Politecnica de Catalunya.
- Kristiansson, B. and Lennartson, B. (2006). Evaluation and simple tuning of PID controllers with high-frequency robustness. *Journal of Process Control*, 16, 91–102.
- López, A.M., Miller, J.A., Smith, C.L., and Murrill, P.W. (1967). Tuning controllers with Error-Integral criteria. *Instrumentation Technology*, 14, 57–62.
- Martin, J., Smith, C.L., and Corripio, A.B. (1975). Controller Tuning from Simple Process Models. *Instrumentation Technology*, 22(12), 39–44.
- Mitchell, M. (1998). *An Introduction to Genetic Algorithms*. MIT Press, USA.
- O'Dwyer, A. (2003). *Handbook of PI and PID Controller Tuning Rules*. Imperial College Press, London, UK.
- Rivera, D.E., Morari, M., and Skogestad, S. (1986). Internal model control 4. pid controller design. *Ind. Eng. Chem. Res.*, 25, 252–265.
- Rovira, A., Murrill, P.W., and Smith, C.L. (1969). Tuning controllers for setpoint changes. *Instrumentation & Control Systems*, 42, 67–69.
- Skogestad, S. (2003). Simple analytic rules for model reduction and PID controller tuning. *Journal of Process Control*, 13, 291–309.
- Tan, W., Liu, J., Chen, T., and Marquez, H.J. (2006). Comparison of some well-known PID tuning formulas. *Computers & Chemical Engineering*, 30, 1416–1423.
- Tavakoli, S., Griffin, I., and Fleming, P.J. (2005). Robust PI controller for load disturbance rejection and setpoint regulation. In *IEEE Conference on Control Applications*. Toronto, Canada.
- Vilanova, R. (2008). IMC based robust PID design: Tuning guidelines and automatic tuning. *Journal of Process Control*, 18.
- Vilanova, R., Alfaro, V.M., Arrieta, O., and Pedret, C. (2010). Analysis of the claimed robustness for PI/PID robust tuning rules. In *18th IEEE Mediterranean Conference on Control and Automation (MED10)*, June 23–25, Marrakech-Morocco.
- Yaniv, O. and Nagurka, M. (2004). Design of PID controllers satisfying gain margin and sensitivity constrains on a set of plants. *Automatica*, 40, 111–116.
- Zhuang, M. and Atherton, D. (1993). Automatic tuning of optimum PID controllers. *IEE Proceedings.Part D.*, 140(3), 216–224.
- Ziegler, J. and Nichols, N. (1942). Optimum Settings for Automatic Controllers. *ASME Transactions*, 759–768.